

Fractional Second Order Sliding Mode Controller for The Control of Wind Generator Based on a DFIG

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Abstract: In order to reduce the chattering phenomena in the conventional sliding mode controller, which appears mostly in the rotor currents of doubly fed induction generator (DFIG), a new Fractional Second Order Sliding Mode Controller is proposed. In this controller, the sliding surfaces are chosen so that they will be compatible with the errors in the rotor currents. The simulation results, show the robustness of the proposed control model, the minimization of the chattering and they improve the total harmonics distortion of the rotor currents.

Keywords: Doubly fed induction generator, Wind turbine, Rotor side converter, Fractional Second Order Sliding Mode controller, Total harmonics distortion,

1. Introduction

In today's human survival and development, energy and environment are the pressing problems to be solved. Wind power, as one source of inexhaustible, clean and pollution-free power, it is prospective of large-scale development and utilization, and is to an effective form by harnessing the wind. At present, the main power generation system of wind driven generator is VSCF (Variable Speed Constant Frequency) wind power generation system for grid.

For the wind energy conversion systems (WECSs) connected to the grid, the doubly fed induction generators received an important consideration compared to other generators types [1-2], due especially to the static converter size, and the wide speed range. Due to several functionalities, the DFIG active and reactive stator powers control is required [1 - 7]. In order to deliver a constant frequency in the DFIG stator, under changing the mechanical speed or wind speed conditions, the rotor of this generator must feeding with variable frequency,

where, the control algorithm law of the rotor side converter (RSC) can be applied as shown in Figure 1.

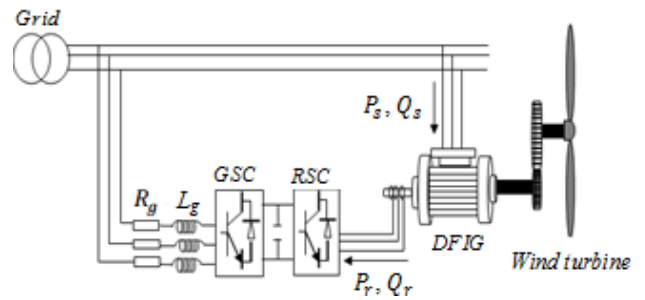


Figure 1: The general structure of a double fed wind generator.

The sliding mode control has received much attention in the field of electrical drives control [8 - 10]. In this method, the response of control system depends only on the sliding surface.

Despite the robustness of this control, the sliding mode control faces the main problem of the chattering phenomenon resulting by the discontinuous control. Various approaches, such as in [11 - 20], have been proposed recently to avoid chattering in the currents, powers and the electromagnetic torque of the electric generators. In this paper, we propose a novel control strategy of the rotor side converter based on the Fractional Second Order Sliding Mode Controller, where, we present another solution to reduce the chattering phenomenon in the conventional sliding mode.

The maximization of the power with speed control, using the proportional integral (PI) regulator, is applied to determinate the optimal values of the reference of the active power used in the stator power control loop.

2. Basic definitions of fractional order calculus:

The two most used definitions for the general

fractional differentiation and integration are the Grunwald-Letnikov (GL) the Riemann-Liouville (RL) definition.

The Grunwald-Letnikov is given by:

$$\alpha D_t^\alpha f(t) = \lim_{h \rightarrow 0} \frac{1}{h^\alpha} \sum_{j=0}^{\lfloor \frac{t-a}{h} \rfloor} (-1)^j \binom{\alpha}{j} f(t-jh) \quad (1)$$

where h is increment

[.] means the integer part and

The Riemann-Liouville (RL) is given by:

$$\alpha D_t^\alpha f(t) = \frac{1}{\Gamma(n-\lambda)} \frac{d^n}{dt^n} \int_a^t \frac{f(\tau)}{(t-\tau)^{\lambda-n+1}} d\tau \quad (2)$$

where $n-1 < \lambda < n$ and $\Gamma(\cdot)$ is the Gamma function .

the Laplace transformation of RL definition for a fractional order λ is given by

$$L\{\alpha D_t^\alpha f(t)\} = s^\lambda F(s) \quad (3)$$

where the $F(s)$ is the Laplace transformation of $f(t)$.

3. Wind Turbine Modeling:

The mechanical power obtained from the wind is given by the following expression [20 – 21]:

$$\begin{aligned} V_{dr} &= R_r \cdot i_{dr} + s \cdot L_r \cdot \sigma \cdot i_{dr} - g \cdot \omega_s \cdot L_r \cdot i_{qr} \cdot \sigma \\ V_{qr} &= R_r \cdot i_{qr} + s \cdot L_r \cdot \sigma \cdot i_{qr} + g \cdot \omega_s \cdot L_r \cdot i_{dr} \cdot \sigma + \frac{g \cdot M \cdot v_s}{L_s} \\ P_a &= C_p \cdot P_{vent} = C_p(\lambda, \beta) \cdot \frac{\rho \cdot \pi \cdot R^2 \cdot V_{vent}^3}{2} \end{aligned} \quad (4)$$

$$\begin{aligned} C_p &= 0.5 - 0.167 \cdot (\beta - 2) \cdot \sin \left[\frac{\pi \cdot (\lambda + 0.1)}{18.5 - 0.3 \cdot \beta} \right] - \\ &0.00184 \cdot (\lambda - 3) \cdot (\beta - 2) \end{aligned} \quad (5)$$

4. Maximum Power Point Tracking

The ($C_p = 0.5$) is reached at tip speed ratio ($\lambda = 9$) and at β constant and equal to two degrees ($\beta = 2^\circ$), the mechanical speed is controlled by the proportional integral controller. Where, the reference rotor speed is given by the following formula:

$$\Omega^* = \frac{\lambda(C_p \max) \cdot \omega_{speed}}{R} \quad (6)$$

5. The Doubly Fed Induction Generator:

The generator dynamics model is given by the following equations system:

$$\begin{cases} v_{ds} = R_s i_{ds} + \frac{d\phi_{ds}}{dt} - \omega_s \phi_{qs} \\ v_{qs} = R_s i_{qs} + \frac{d\phi_{qs}}{dt} + \omega_s \phi_{ds} \\ v_{dr} = R_r i_{dr} + \frac{d\phi_{dr}}{dt} - (\omega_s - \omega) \phi_{qr} \\ v_{qr} = R_r i_{qr} + \frac{d\phi_{qr}}{dt} + (\omega_s - \omega) \phi_{dr} \end{cases} \quad (7)$$

Where

$$\begin{cases} \phi_{ds} = L_s \cdot i_{ds} + M \cdot i_{dr} \\ \phi_{qs} = L_s \cdot i_{qs} + M i_{qr} \\ \phi_{dr} = L_r \cdot i_{dr} + M \cdot i_{ds} \\ \phi_{qr} = L_r \cdot i_{qr} + M \cdot i_{qs} \end{cases} \quad (8)$$

And the mechanical model of the doubly fed induction generator will be given by the following expression:

$$C_e - C_r = J \frac{d\Omega}{dt} + f\Omega \quad (9)$$

6. A Fractional Second Order Sliding Mode Controller Design:

The fractional order sliding surfaces are:

$$S(i_{dr}) = e_1 + c_d \cdot D^{-1} \cdot e_1 \quad (10)$$

$$S(i_{qr}) = e_2 + c_q \cdot D^{-1} \cdot e_2 \quad (11)$$

$$S = [S(i_{dr}) \quad S(i_{qr})]$$

We define the rotor currents errors, respectively, as follow: $e_1 = (i_{dr}^* - i_{dr})$ and $e_2 = (i_{qr}^* - i_{qr})$, where the i_{dr}^* and i_{qr}^* are, respectively.

where the rotor volage can be written as follow:

The nonlinear state space model is given under the form:

$$\dot{x} = f(x) + gV_r \quad (12)$$

where

$$V_r = [V_{dr} \quad V_{qr}]^T$$

$$x = [i_{dr} \quad i_{qr}]^T$$

where α represents order of fractional operator, c_d , c_q are positive tuning parameters.

The main condition in the sliding mode is achieving the following equality:

$$\frac{dS(i_{dr})}{dt} = \frac{dS(i_{qr})}{dt} = 0$$

So:

$$\dot{S} = \begin{bmatrix} S(i_{dr}) \\ S(i_{qr}) \end{bmatrix} = \begin{bmatrix} \dot{e}_1 + c_d \cdot D^{1-\alpha} \cdot e_1 \\ \dot{e}_2 + c_q \cdot D^{1-\alpha} \cdot e_2 \end{bmatrix} \quad (13)$$

And the rotor controlled voltage may be derived according to the control law:

$$V_{dr} = V_{dr ST} + V_{dr eq} \quad (14)$$

$$V_{qr} = V_{qr ST} + V_{qr eq} \quad (15)$$

$$V_{eq}^{d,q} = g^{-1} \left[-f(x) + x^* - k \cdot \text{sgn} - \begin{bmatrix} c_d \cdot D^{1-\alpha} \cdot e_1 \\ c_q \cdot D^{1-\alpha} \cdot e_2 \end{bmatrix} \right]$$

where

$k = [k_1 \quad k_2]^T$ are the positive control gains.

✓ **The super- twisting algorithm:**

The super-twisting algorithm (STA), has been adopted given by:

$$\begin{cases} STA(e_1) = \delta_q |e_1|^{(1/2)} \text{sgn}(e_1) - w_q \int \text{sgn}(e_1) dt \\ STA(e_2) = \delta_d |e_2|^{(1/2)} \text{sgn}(e_2) - w_d \int \text{sgn}(e_2) dt \end{cases} \quad (16)$$

where δ_d , δ_q , w_d and w_q are the positive control gains, The function sgn , equations(14) and (15), will be substituted by a STA,

7. Simulation Studies

Results of simulation studies, performed using MATLAB software, in response to the change of wind speed as shown in Figure 2, are described in this Section.

The DFIG - wind turbine system parameters are given Table 1.

Table I: The DFIG - wind turbine system parameters.

Parameters	Units	Value
The nominal power	[MW]	1.5
The stator voltage	[V]	690
The stator frequency	[Hz]	50
The number of poles pairs	[]	2
The stator resistance	[Ω]	0.012
The rotor resistance	[Ω]	0.021
The stator inductance	[H]	0.0137
The rotor inductance	[H]	0.0136
The mutual inductance	[H]	0.0135
The inertia	[Kg.m ²]	1000
The friction coefficient	[N.m.s/rad]	0.0024
The tip speed ratio max	[]	9
The power coefficient $C_{P \max}$	[]	0.5
The radius of the wind	[m]	35.25
The gain multiplier	[]	90
The air density	[kg/m ³]	1.225

The DFIG- wind turbine system tracks the maximum

power point as the wind speed changes. It is confirmed from Figure 3 where the power coefficient follows its optimal reference and is equal to 0.5. Similarly, the BDFIG control system ensures that the optimal constant tip speed ratio value, equal to 9, is maintained as shown in Figure 4. It is evident from Figure 5 that the mechanical speed tracks its optimal reference almost perfectly.

Response of the DFIG, Figures 6 and 7, shows that the active and reactive powers reach their optimally values. Neglecting the starting transients in the first 0.5 s, a good dynamic performance is clearly achieved as there is minimum overshoot in the powers around their desired values in the transitory regime and the response time is practically negligible. It is observed from Figures 8 and 9 that the two components of the rotor currents reach their reference values in a negligible response time and a low overshoot. It is also seen from Figure 9 that the quadrature components of current keep track of the wind speed profile, while it is clearly in the same figure that the optimal values of the quadrature component of the rotor currents is obtained through the controlling of the DFIG- wind turbine system under the maximizing the power in the system.

Total harmonic distortion of the rotor currents with Fractional Second Order Sliding Mode Controller is proposed is illustrated in Figure 11. THD is equal to 1.4% indicating that the harmonic content of the rotor currents is very small. It is also seen from the rotor currents that there is no chattering phenomenon. The DFIG electromagnetic torque, depicted in Figure 12, is the image of the wind profile.

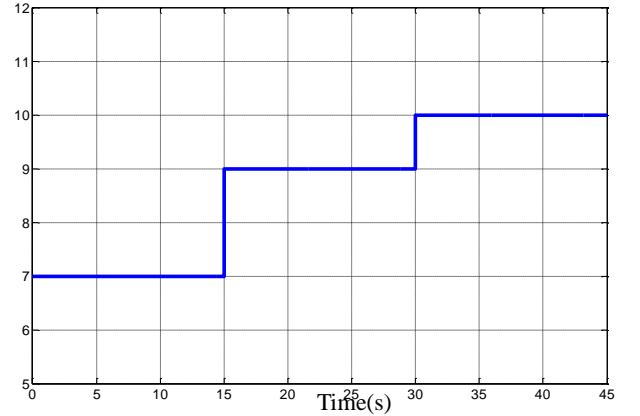


Figure .2 : The wind speed applied to wind turbine.

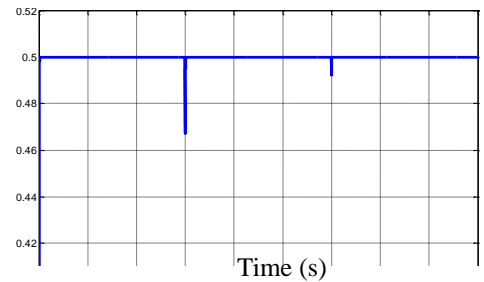


Figure .3 : The power coefficient C_p .

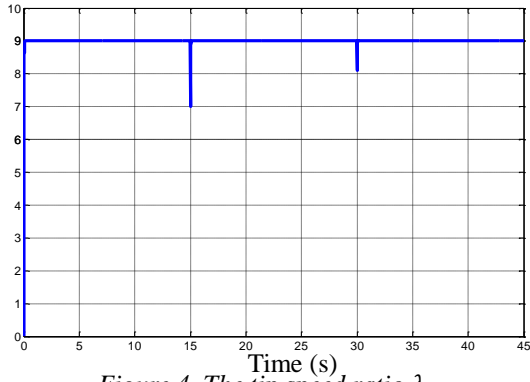


Figure 4. The tip speed ratio λ .

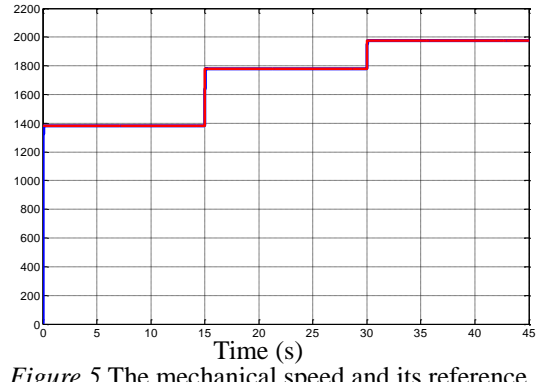


Figure.5 The mechanical speed and its reference (rpm/min).

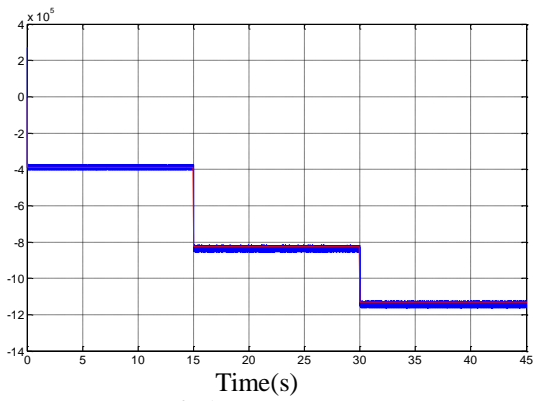


Figure.6 The stator active power[W]

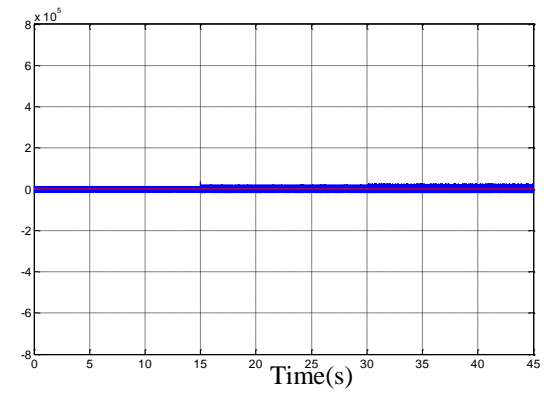


Figure.7 The stator active power[VAR].

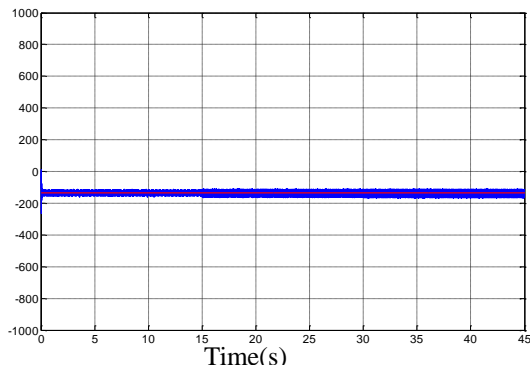


Figure.8. The direct component of the rotor currents[A].

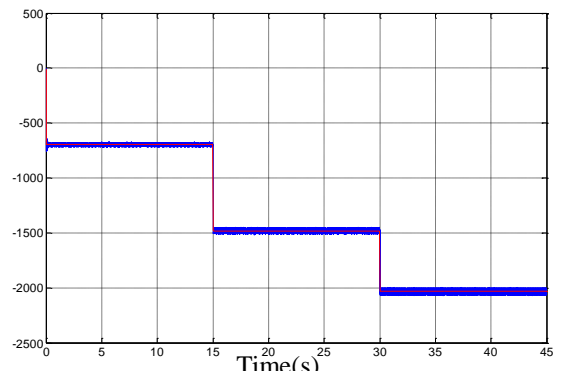


Figure.9. The quadrature component of the rotor currents[A].

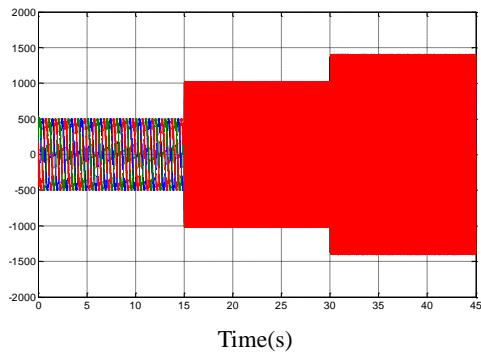
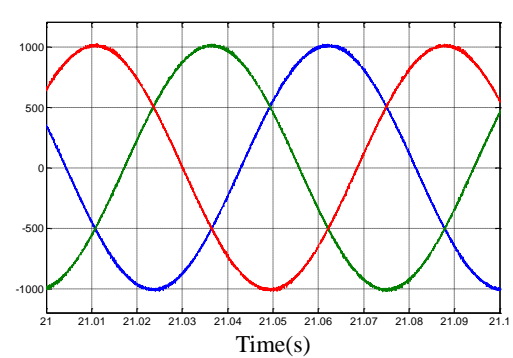
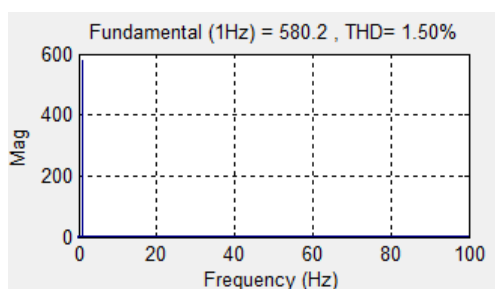
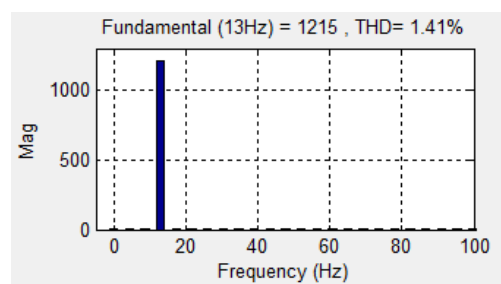


Figure.10 the rotor currents with his zoom.





(a) with $v = 7m/s$



(b) with $v = 9m/s$

Figure.11 Total harmonic distortion of the rotor currents.

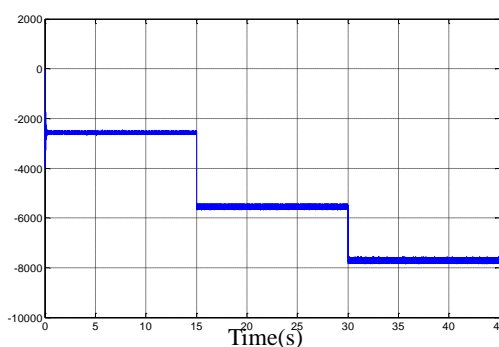


Figure 12 : The electromagnetic torque.

8. Conclusion:

A fractional second order sliding mode controller for indirect power control of a DFIG - wind turbine system is proposed in this work. As a consequence, the chattering phenomenon, which appears in the generator torque and currents, is avoided. The simulation results given by the proposed model show that the stator active and reactive powers follow perfectly the reference values with zero steady state error.

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